

# Dynamics, Synchronization and SPICE Implementation of a Memristive System with Hidden Hyperchaotic Attractor

Viet-Thanh Pham, Sundarapandian Vaidyanathan, Christos K. Volos,  
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**Abstract** The realization of memristor in nanoscale size has received considerable attention recently because memristor can be applied in different potential areas such as spiking neural network, high-speed computing, synapses of biological systems, flexible circuits, nonvolatile memory, artificial intelligence, modeling of complex systems or low power devices and sensing. Interestingly, memristor has been used as a nonlinear element to generate chaos in memristive system. In this chapter, a new memristive system is proposed. The fundamental dynamics properties of such memristive system are discovered through equilibria, Lyapunov exponents, and Kaplan–York dimension. Especially, hidden attractor and hyperchaos can be observed in this new system. Moreover, synchronization for such system is studied and simulation results are presented showing the accuracy of the introduced synchronization scheme. An electronic circuit modelling such hyperchaotic memristive system is also reported to verify its feasibility.

**Keywords** Chaos · Hyperchaos · Lyapunov exponents · Hidden attractor · No-equilibrium · Memristor · Synchronization · Circuit · SPICE

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## 1 Introduction

After the discovery of Lorenz's model for atmospheric convection [1], there has been significant interest in chaotic systems [2–9]. In the past few decades, different chaotic systems have been reported such as Rössler system [10], Arneodo system [11], Chen system [6], Lü system [12], Vaidyanathan system [13], time-delay systems [14] and so on [15, 16]. Chaotic behaviors are useful and have been applied in many fields, for example a double-scroll chaotic attractor has been used to generate true random bits [17], chaotic path planning has been generated for autonomous mobile robots [18], fingerprint images encryption scheme based on chaotic attractors has been implemented, or applications of time delay systems in secure communication have been proposed [19] due to their complex dynamics.

In addition, hyperchaotic system was introduced and studied [20]. Hyperchaotic system is characterized by more than one positive Lyapunov exponent and, thus, presents a higher level of complexity with respect to chaotic system [21]. As a result, hyperchaos is better than conventional chaos in a variety of areas, for instance, hyperchaos increases the security of chaotic-based communication systems significantly [22, 23]. Moreover hyperchaos has been used in diverse applications such as cryptosystems [24], neural networks [25], secure communications [22, 23], or laser design [26]. Especially, the intrinsic nonlinear characteristic of memristor has been exploited in designing hyperchaotic oscillators. Some recent researches show that memristor is a potential candidate for generating hyperchaos [27, 28].

In this chapter, our work introduces a memristive system which can exhibit hyperchaotic attractors. Moreover such memristive system does not have equilibrium points. This chapter is organized as follows. Section 2 summarizes related works. Section 3 gives a brief representation to the memristive system. Dynamics and properties of such memristive system is introduced in Sect. 4 while the adaptive synchronization scheme is studied in Sect. 5. Section 6 presents circuit implementation of memristive system using SPICE. Finally, conclusions are drawn in Sect. 7.

## 2 Related Work

Motivated by complex dynamical behaviors of hyperchaotic systems and special features of memristor, some memristor-based hyperchaotic systems have been introduced, recently. Hyperchaos was generated by combining a memristor with cubic nonlinear characteristics and a modified canonical Chua's circuit [28]. This memristor-based modified canonical Chua's circuit is a five-dimensional hyperchaotic oscillator. By extending the HP memristor-based canonical Chua's oscillator, a six-dimensional hyperchaotic oscillator was designed [29]. Authors used a configuration based on two HP memristors in antiparallel [27]. Four-dimensional hyperchaotic memristive systems were discovered by Li et al. [30, 31]. A 4D memristive system with a line of equilibrium was presented in [30] while another memristive system with an

uncountable infinite number of stable and unstable equilibria was reported in [31]. A memristor-based hyperchaotic system without equilibrium was introduced in [32]. These memristive systems belong to a new category of chaotic systems with hidden attractors [33, 34].

The terminology “hidden attractor” has been introduced recently although the fact that the problem of analyzing hidden oscillations and to the finding of hidden oscillations in automatic control systems were studied a long time ago. According to a new classification of chaotic dynamics proposed by Leonov and Kuznetsov [33–35], there are two types of attractors: self-excited attractors and hidden attractors. A self-excited attractor has a basin of attraction that is excited from unstable equilibria. In contrast, hidden attractor cannot be found by using a numerical method in which a trajectory started from a point on the unstable manifold in the neighbourhood of an unstable equilibrium [35]. The discovery of dynamical systems with hidden attractors is a great challenge due to their appearance in many research fields such as in mechanics, secure communication and electronics [34, 36–39]. For example, hidden attractor in smooth Chua’s system was reported in [40]. Hidden oscillations in mathematical model of drilling system [41] and hidden oscillations in nonlinear control systems [42] were witnessed. Various examples of hidden attractors were summarized in [43–46]. Hidden attractors were observed in a 4-D Rikitake dynamo system [47] or 5-D hyperchaotic Rikitake dynamo system [48]. Hidden attractors in a chaotic system with an exponential nonlinear term were introduced in [49]. In addition, algorithms for searching for hidden oscillations were presented in [50, 51].

Motivated by complex dynamical behaviors of chaotic systems, noticeable characteristics of memristor, and unknown features of hidden attractors, studying memristive hyperchaotic systems with hidden attractors is still an attractive research direction [30, 31].

### 3 Model of the Memristive System

A flux-controlled memristor is considered in this work. Its memductance is a second degree polynomial function:

$$W(\varphi) = \alpha + 3\beta\varphi^2, \quad (1)$$

with  $\alpha = 0.4$  and  $\beta = 0.001$ . Memductance (1) is similar to known memductance [28, 30, 52, 53]. Using this memristor, a four-dimensional memristive system is proposed as

$$\begin{cases} \dot{x} = 36y - 36x \\ \dot{y} = -2xz + 20y - axW(\varphi) - b \\ \dot{z} = 2xy - 3z \\ \dot{\varphi} = 2x, \end{cases} \quad (2)$$

where  $a, b$  are parameters, and  $W(\varphi)$  is the memductance as introduced in (1). It is noting that the memristive system (2) has an uncountable number of equilibrium points when  $b = 0$ . Moreover, system (2) generate hyperchaos for different values of the parameter  $a$ . For example, hyperchaotic attractors is obtained when  $a = 30, b = 0$  and the chosen initial conditions are  $(x(0), y(0), z(0), \varphi(0)) = (0.5, 0, 0.5, 0)$ . In this case, memristive system (2) is similar to the reported one in [31], therefore it will be not discussed in the next sections.

## 4 Dynamics and Properties of the Memristive System

The memristive system (2) is investigated when  $b \neq 0$ . It is easy to obtain the equilibrium points for system (2) by solving  $\dot{x} = 0, \dot{y} = 0, \dot{z} = 0$ , and  $\dot{\varphi} = 0$  that is

$$36y - 36x = 0, \quad (3)$$

$$-2xz + 20y - axW(\varphi) - b = 0, \quad (4)$$

$$2xy - 3z = 0, \quad (5)$$

$$2x = 0, \quad (6)$$

From (3), (5) and (6), we have  $x = y = z = 0$ . As a results, Eq. (4) reduces to  $b = 0$ , which is an contradiction. Hence there are not equilibrium points in memristive system (2).

In this work, the parameters are selected as  $a = 30, b = 0.001$  and the initial conditions are

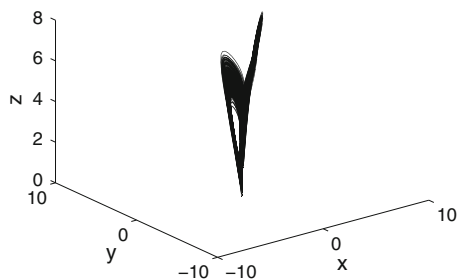
$$(x(0), y(0), z(0), \varphi(0)) = (0.5, 0, 0.5, 0). \quad (7)$$

Lyapunov exponents, which measure the exponential rates of the divergence and convergence of nearby trajectories in the phase space of the chaotic system [8, 54], are calculated using the well-known algorithm in [55]. The Lyapunov exponents of the system (2) are

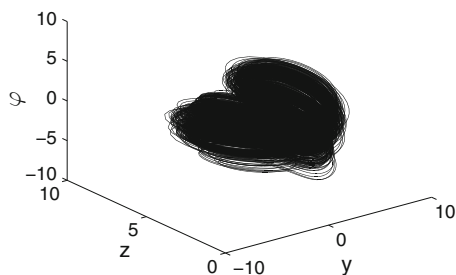
$$\lambda_1 = 0.2590, \lambda_2 = 0.0658, \lambda_3 = 0, \lambda_4 = -19.3246. \quad (8)$$

It is noting that the sum of the Lyapunov exponents is negative, and so the novel memristive hyperchaotic system is dissipative. There are two positive Lyapunov exponents, one zero and one negative Lyapunov exponents. Thus, the memristive system (2) is a four-dimension hyperchaotic system according to [20]. It is worth noting that this memristive system can be classified as a hyperchaotic system with hidden strange attractor because its basin of attractor does not contain neighbourhoods of equilibria [33, 34]. The 3-D and 2-D projections of the hyperchaotic attractors without equilibrium in this case are illustrated in Figs. 1, 2, 3, 4, 5 and 6.

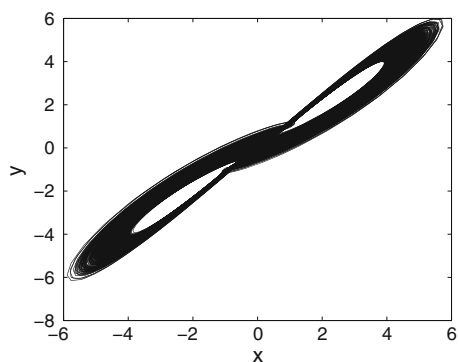
**Fig. 1** 3-D projection of the hyperchaotic memristive system without equilibrium (2) in the  $(x, y, z)$ -space



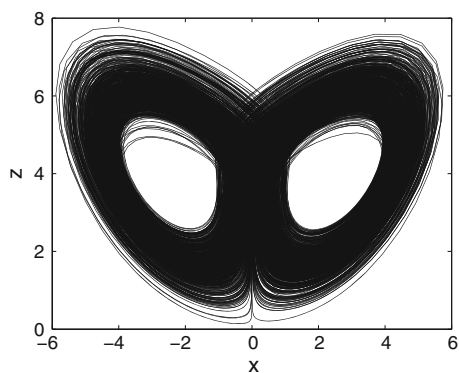
**Fig. 2** 3-D projection of the hyperchaotic memristive system without equilibrium (2) in the  $(y, z, \varphi)$ -space



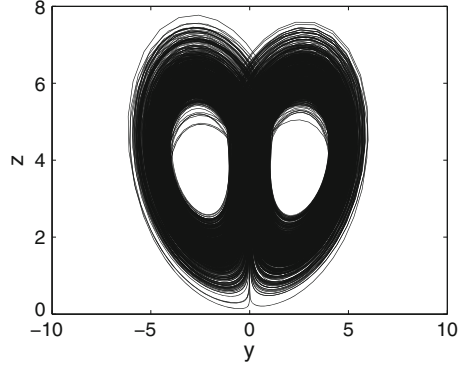
**Fig. 3** 2-D projection of the hyperchaotic memristive system without equilibrium (2) in the  $(x, y)$ -plane



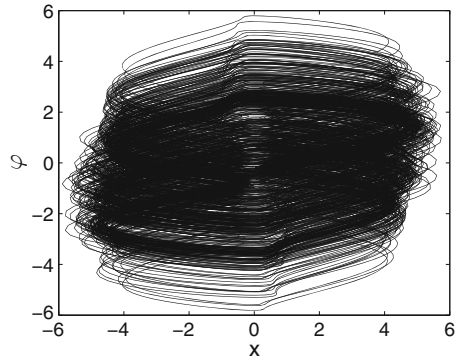
**Fig. 4** 2-D projection of the hyperchaotic memristive system without equilibrium (2) in the  $(x, z)$ -plane



**Fig. 5** 2-D projection of the hyperchaotic memristive system without equilibrium (2) in the  $(y, z)$ -plane



**Fig. 6** 2-D projection of the hyperchaotic memristive system without equilibrium (2) in the  $(x, \varphi)$ -plane



It is known that the Kaplan–Yorke fractional dimension, which presents the complexity of attractor [56], is defined by

$$D_{KY} = j + \frac{1}{|\lambda_{j+1}|} \sum_{i=1}^j \lambda_i, \quad (9)$$

where  $j$  is the largest integer satisfying  $\sum_{i=1}^j \lambda_i \geq 0$  and  $\sum_{i=1}^{j+1} \lambda_i < 0$ . The calculated fractional dimension of memristive system (2) when  $a = 30$ ,  $b = 0.001$  is

$$D_{KY} = 3 + \frac{\lambda_1 + \lambda_2 + \lambda_3}{|\lambda_4|} = 3.0168. \quad (10)$$

Equation (10) indicates a strange attractor.

## 5 Adaptive Anti-synchronization of the Memristive System

The possibility of synchronization of two coupled chaotic systems [57–60] is one of the most vital characteristics relating to chaotic systems and their applications. Different research activities using synchronization of nonlinear systems have been investigated in literature [38, 61–69]. For instant, synchronized states in a ring of mutually coupled self-sustained nonlinear electrical oscillators were considered in [70], ragged synchronizability of coupled oscillators was observed in [71], various synchronization phenomena in bidirectionally coupled double-scroll circuits were reported in [72], or observer for synchronization of chaotic systems with application to secure data transmission was studied in [73]. Many synchronization schemes have been introduced such as lag synchronization [74], frequency synchronization [75], projective-anticipating synchronization [76], anti-synchronization [77], adaptive synchronization [78], or hybrid chaos synchronization [63], etc. Here we consider the adaptive synchronization of identical memristive hyperchaotic systems with two unknown parameters.

In this section, we consider the memristive system (2) as the master system as follows

$$\begin{cases} \dot{x}_1 = 36y_1 - 36x_1 \\ \dot{y}_1 = -2x_1z_1 + 20y_1 - ax_1W(\varphi_1) - b \\ \dot{z}_1 = 2x_1y_1 - 3z_1 \\ \dot{\varphi}_1 = 2x_1. \end{cases} \quad (11)$$

The states of the master system (11) are  $x_1, y_1, z_1, \varphi_1$ , and  $W(\varphi_1)$  is the memductance as given in (1). The slave system is considered as the controlled memristive system and its dynamics is given as

$$\begin{cases} \dot{x}_2 = 36y_2 - 36x_2 + u_x \\ \dot{y}_2 = -2x_2z_2 + 20y_2 - ax_2W(\varphi_2) - b + u_y \\ \dot{z}_2 = 2x_2y_2 - 3z_2 + u_z \\ \dot{\varphi}_2 = 2x_2 + u_\varphi, \end{cases} \quad (12)$$

where  $x_2, y_2, z_2, \varphi_2$ , are the states of the slave system while  $u_x, u_y, u_z, u_\varphi$  are the adaptive controls. These controls will be constructed for the anti-synchronization of the master and slave systems. In order to estimate unknown parameters  $a$  and  $b$ ,  $A(t)$  and  $B(t)$  are used.

The anti-synchronization error between memristive systems (11) and (12) is described by the following relation

$$\begin{cases} e_x = x_1 + x_2 \\ e_y = y_1 + y_2 \\ e_z = z_1 + z_2 \\ e_\varphi = \varphi_1 + \varphi_2. \end{cases} \quad (13)$$

Therefore, the anti-synchronization error dynamics is determined by

$$\begin{cases} \dot{e}_x = 36e_x - 36e_y + u_x \\ \dot{e}_y = -2(x_1z_1 + x_2z_2) + 20e_y - a(x_1W(\varphi_1) + x_2W(\varphi_2)) - 2b + u_y \\ \dot{e}_z = 2(x_1y_1 + x_2y_2) - 3e_z + u_z \\ \dot{e}_\varphi = 2e_x + u_\varphi. \end{cases} \quad (14)$$

Our goal is to find the appropriate controllers  $u_x$ ,  $u_y$ ,  $u_z$ ,  $u_\varphi$  to stabilize the system (14). Thus, we propose the following controllers for system (14):

$$\begin{cases} u_x = -36e_x + 36e_y - k_x e_x \\ u_y = 2(x_1z_1 + x_2z_2) - 20e_y + A(t)(x_1W(\varphi_1) + x_2W(\varphi_2)) + 2B(t) - k_y e_y \\ u_z = -2(x_1y_1 + x_2y_2) + 3e_z - k_z e_z \\ u_\varphi = -2e_x - k_\varphi e_\varphi, \end{cases} \quad (15)$$

where  $k_x$ ,  $k_y$ ,  $k_z$ ,  $k_\varphi$  are positive gain constants for each controllers and  $A(t)$ ,  $B(t)$  are the estimate values for unknown system parameters. The update laws for the unknown parameters are defined as

$$\begin{cases} \dot{A} = -e_y(x_1W(\varphi_1) + x_2W(\varphi_2)) \\ \dot{B} = -2e_y. \end{cases} \quad (16)$$

Next, the main result of this section will be presented and proved.

**Theorem 5.1** *If the adaptive controller (15) and the updating laws of parameter (16) are chosen, the anti-synchronization between the master system (11) and the slave system (12) is achieved.*

*Proof* Here  $e_a(t)$  and  $e_b(t)$  are the parameter estimation errors given as

$$\begin{cases} e_a(t) = a - A(t) \\ e_b(t) = b - B(t). \end{cases} \quad (17)$$

Differentiating (17) with respect to  $t$ , we obtain

$$\begin{cases} \dot{e}_a(t) = -\dot{A}(t) \\ \dot{e}_b(t) = -\dot{B}(t). \end{cases} \quad (18)$$

Substituting adaptive control law (15) into (14), the closed-loop error dynamics is determined as

$$\begin{cases} \dot{e}_x = -k_x e_x \\ \dot{e}_y = -(a - A(t))(x_1W(\varphi_1) + x_2W(\varphi_2)) - 2(b - B(t)) - k_y e_y \\ \dot{e}_z = -k_z e_z \\ \dot{e}_\varphi = -k_\varphi e_\varphi \end{cases} \quad (19)$$



Then substituting (17) into (19), we have

$$\begin{cases} \dot{e}_x = -k_x e_x \\ \dot{e}_y = -e_a (x_1 W(\varphi_1) + x_2 W(\varphi_2)) - 2e_b - k_y e_y \\ \dot{e}_z = -k_z e_z \\ \dot{e}_\varphi = -k_\varphi e_\varphi \end{cases} \quad (20)$$

We consider the Lyapunov function as

$$\begin{aligned} V(t) &= V(e_x, e_y, e_z, e_\varphi, e_a, e_b) \\ &= \frac{1}{2} (e_x^2 + e_y^2 + e_z^2 + e_\varphi^2 + e_a^2 + e_b^2). \end{aligned} \quad (21)$$

The Lyapunov function is clearly definite positive.

Taking time derivative of (21) along the trajectories of (13) and (17) we get

$$\dot{V}(t) = e_x \dot{e}_x + e_y \dot{e}_y + e_z \dot{e}_z + e_\varphi \dot{e}_\varphi + e_a \dot{e}_a + e_b \dot{e}_b. \quad (22)$$

From (18), (20), and (22) we have

$$\begin{aligned} \dot{V}(t) &= -k_x e_x^2 - e_a [e_y (x_1 W(\varphi_1) + x_2 W(\varphi_2)) + \dot{A}] \\ &\quad - e_b (2e_y + \dot{B}) - k_y e_y^2 - k_z e_z^2 - k_\varphi e_\varphi^2. \end{aligned} \quad (23)$$

Then by applying the parameter update law (16), Eq. (23) become

$$\dot{V}(t) = -k_x e_x^2 - k_y e_y^2 - k_z e_z^2 - k_\varphi e_\varphi^2. \quad (24)$$

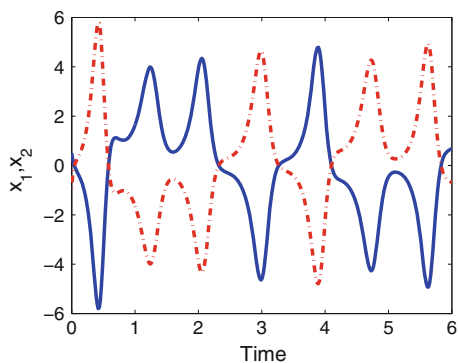
Obviously, derivative of the Lyapunov function is definite negative. According to the Lyapunov stability [79, 80] we obtain  $e_x(t) \rightarrow 0$ ,  $e_y(t) \rightarrow 0$ ,  $e_z(t) \rightarrow 0$ ,  $e_\varphi(t) \rightarrow 0$ ,  $e_a(t) \rightarrow 0$ ,  $e_b(t) \rightarrow 0$  exponentially when  $t \rightarrow 0$  that is, anti-synchronization between master and slave system. This completes the proof.  $\square$

We illustrate the proposed anti-synchronization scheme with a numerical example. In the numerical simulations, the fourth-order Runge–Kutta method is used to solve the systems. The parameters of the memristive hyperchaotic systems are selected as  $a = 30$ ,  $b = 0.001$  and the positive gain constant as  $k = 4$ . The initial conditions of the master system (11) and the slave system (12) have been chosen as  $x_1(0) = 0.5$ ,  $y_1(0) = 0$ ,  $z_1(0) = 0.5$ ,  $\varphi_1(0) = 0$  and  $x_2(0) = -0.9$ ,  $y_2(0) = -0.4$ ,  $z_2(0) = 0.8$ ,  $\varphi_2(0) = 0.5$ , respectively. We assumed that the initial values of the parameter estimates are  $A(0) = 29$  and  $B(0) = 0.5$ .

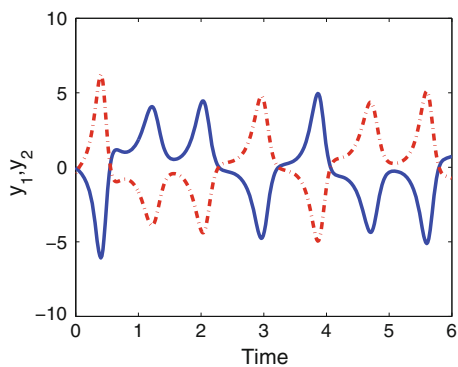
It is easy to see that when adaptive control law (15) and the update law for the parameter estimates (16) are applied, the anti-synchronization of the master (11) and slave system (12) occurred as illustrated in Figs. 7, 8, 9 and 10. It is noting that time series of master states are denoted as blue solid lines while corresponding slave states are plotted as red dash-dot lines in such figures. In addition, the time-history of the anti-synchronization errors  $e_x$ ,  $e_y$ ,  $e_z$ , and  $e_\varphi$  is presented in Fig. 11.

The anti-synchronization errors converge to the zero, which implies that the chaos anti-synchronization between the memristive systems is realized.

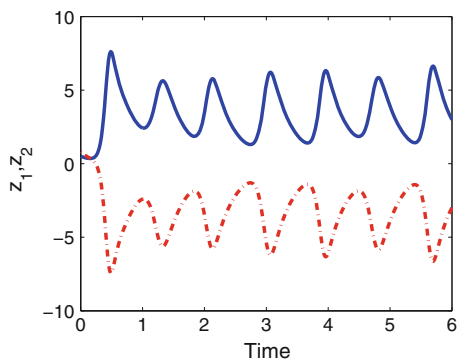
**Fig. 7** Anti-synchronization of the states  $x_1(t)$  and  $x_2(t)$



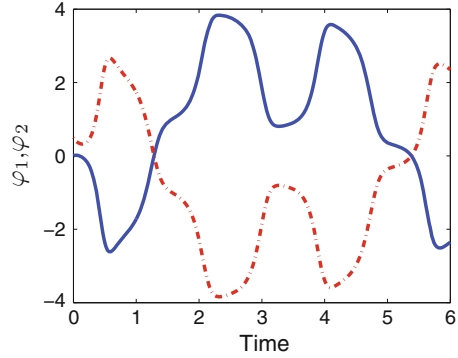
**Fig. 8** Anti-synchronization of the states  $y_1(t)$  and  $y_2(t)$



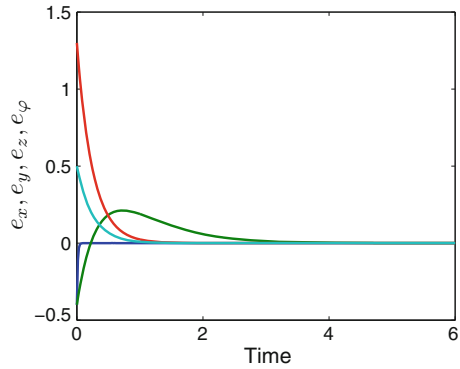
**Fig. 9** Anti-synchronization of the states  $z_1(t)$  and  $z_2(t)$



**Fig. 10** Anti-synchronization of the states  $\varphi_1(t)$  and  $\varphi_2(t)$



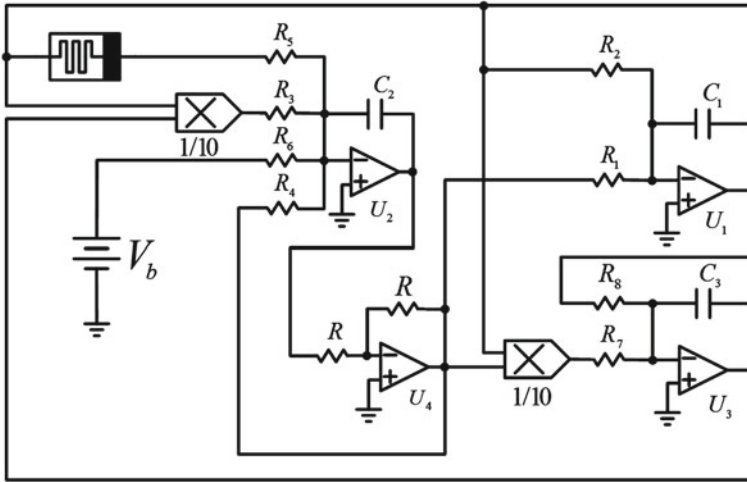
**Fig. 11** Time series of the anti-synchronization errors  $e_x, e_y, e_z$ , and  $e_\varphi$



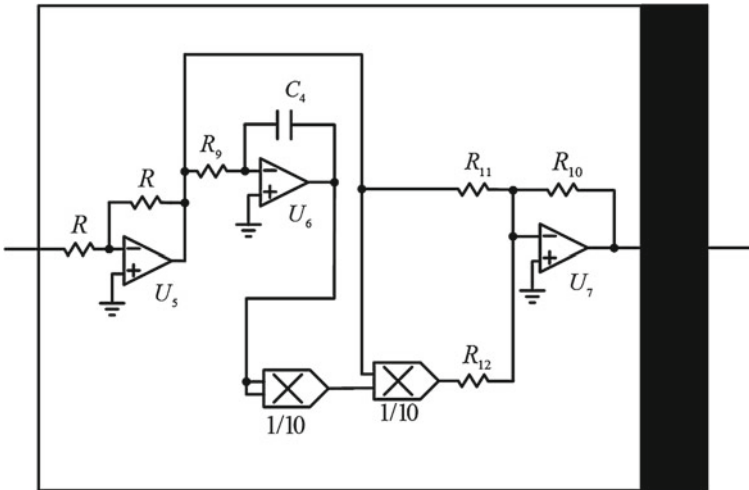
## 6 Circuit Realization of the Memristive System

Circuitual design of chaotic/hyperchaotic systems plays an important role on the field of nonlinear science due to its applications in secure communication, signal processing, random bit generator, or path planning for autonomous mobile robot etc. [17, 18, 22, 62, 81, 82]. In addition, circuitual implementation of chaotic/hyperchaotic systems is also provide an effective approach for investigating dynamics of such theoretical models [61, 83]. For example, chaotic attractors can be observed on the oscilloscope easily or experimental bifurcation diagrams can be obtained by varying the values of variable resistors [84, 85].

Therefore, in this work, an electronic circuit is introduced to implement memristive system (2). By using the operational amplifiers approach [85], the circuit is proposed as shown in Fig. 12. Here the variables  $x, y, z, \varphi$  of memristive system (2) are the voltages across the capacitor  $C_1, C_2, C_3$ , and  $C_4$ , respectively. It is noted



**Fig. 12** Schematic of the circuit which modelling hyperchaotic system (2) with the presence of the memristor



**Fig. 13** Schematic of the circuit emulating the memristor

that the detailed schematic of the memristor in Fig. 12 is presented in Fig. 13. This sub-circuit of memristor emulates the memristive device only due to the fact that there are not any commercial off-the-shelf memristive device in the market at the moment [86]. The corresponding circuitual equations of circuit can be described as

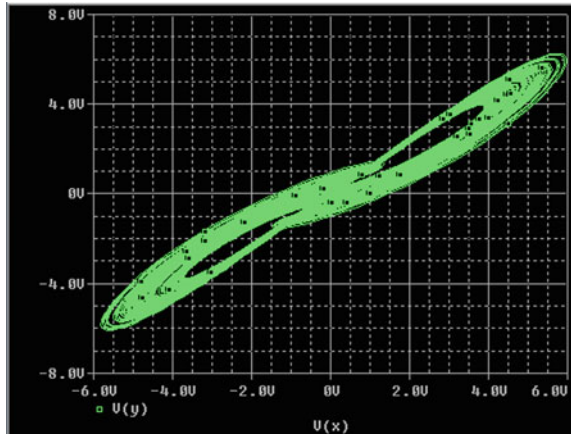
$$\begin{cases} \frac{dv_{C_1}}{dt} = \frac{1}{R_1 C_1} v_{C_2} - \frac{1}{R_2 C_1} v_{C_1} \\ \frac{dv_{C_2}}{dt} = -\frac{1}{10 R_3 C_2} v_{C_1} v_{C_3} + \frac{1}{R_4 C_2} v_{C_2} - \frac{1}{R_6 C_2} V_b \\ \quad - \frac{1}{R_5 C_2} v_{C_1} \left( \frac{R_{10}}{R_{11}} + \frac{R_{10}}{100 R_{12}} v_{C_4}^2 \right) \\ \frac{dv_{C_3}}{dt} = \frac{1}{10 R_7 C_3} v_{C_1} v_{C_2} - \frac{1}{R_8 C_3} v_{C_3} \\ \frac{dv_{C_4}}{dt} = \frac{1}{R_9 C_4} v_{C_1}, \end{cases} \quad (25)$$

where  $v_{C_1}$ ,  $v_{C_2}$ ,  $v_{C_3}$ , and  $v_{C_4}$  are the voltages across the capacitors  $C_1$ ,  $C_2$ ,  $C_3$ , and  $C_4$ , respectively.

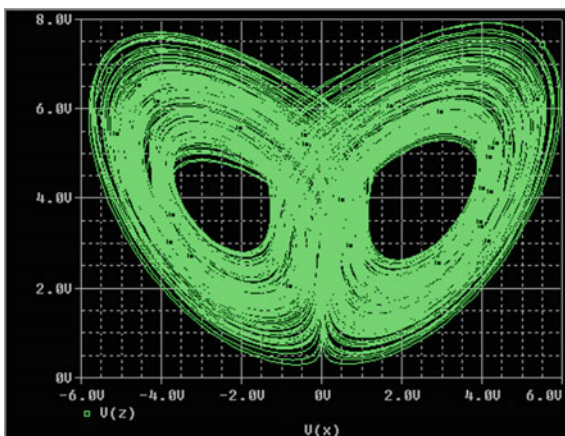
The power supplies of all active devices are  $\pm 15V_{DC}$  and the operational amplifiers TL084 are used in this work. The values of components in Figs. 12 and 13 are chosen as follows:  $R_1 = R_2 = 1 \text{ k}\Omega$ ,  $R_3 = R_4 = R_7 = 1.8 \text{ k}\Omega$ ,  $R_5 = 1.2 \text{ k}\Omega$ ,  $R_6 = 3.6 \text{ M}\Omega$ ,  $R_8 = 12 \text{ k}\Omega$ ,  $R_9 = 18 \text{ k}\Omega$ ,  $R_{10} = R = 36 \text{ k}\Omega$ ,  $R_{11} = 90 \text{ k}\Omega$ ,  $R_{12} = 120 \text{ k}\Omega$ ,  $V_b = 0.1 V_{DC}$ , and  $C_1 = C_2 = C_3 = C_4 = 4.7 \text{ nF}$ .

The designed circuit is implemented in SPICE. The obtained results are displayed in Figs. 14, 15, 16 and 17 which show the hyperchaotic attractors of the designed circuit in different phase planes  $(v_{C_1}, v_{C_2})$ ,  $(v_{C_1}, v_{C_3})$ ,  $(v_{C_2}, v_{C_3})$ , and  $(v_{C_1}, v_{C_4})$ , respectively. Theoretical attractors (see Figs. 3, 4, 5 and 6) are similar with the circuital ones (see Figs. 14, 15, 16 and 17). Moreover, the designed circuit confirms the feasibility of the memristive system.

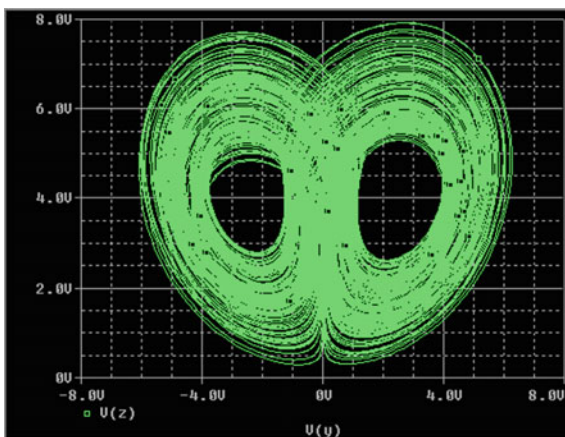
**Fig. 14** Hyperchaotic attractor of the designed circuit obtained from SPICE in the  $(v_{C_1}, v_{C_2})$  phase plane



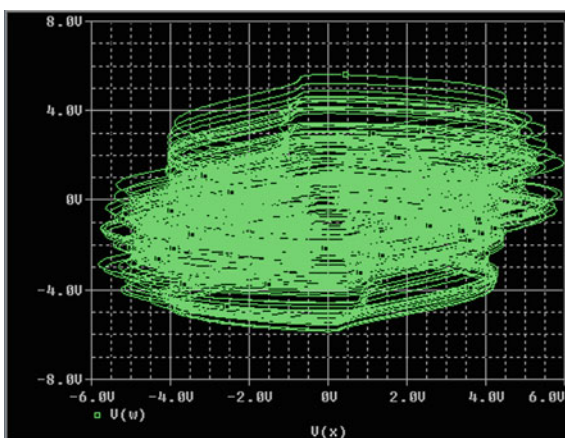
**Fig. 15** Hyperchaotic attractor of the designed circuit obtained from SPICE in the  $(v_{C_1}, v_{C_3})$  phase plane



**Fig. 16** Hyperchaotic attractor of the designed circuit obtained from SPICE in the  $(v_{C_2}, v_{C_3})$  phase plane



**Fig. 17** Hyperchaotic attractor of the designed circuit obtained from SPICE in the  $(v_{C_1}, v_{C_4})$  phase plane



## 7 Conclusion

A memristive system, which is built by using a memristor, is proposed in this work. The presence of the memristor creates the special features of such as hyperchaos, the absence of equilibrium points, and hidden attractors. Fundamental dynamical behaviors of the memristive hyperchaotic system are investigated through calculating equilibrium points, phase portraits of chaotic attractors, Lyapunov exponents and Kaplan–Yorke dimension. In addition, the capacity of synchronization of memristive systems and the feasibility of such memristive system without equilibrium are verified through anti-synchronization scheme and circuital implementation, respectively.

It is worth noting that the memristive system can exhibit double-scroll hyperchaotic attractor despite the equilibrium points have disappeared. It has been known that equilibrium points of a dynamical system, especially a chaotic one, play an important role when generating multi-scroll attractors. Therefore, investigating no-equilibrium memristive systems with multi-scroll attractors will be studied in future works.

Moreover, the memristive system has potential applications in secure communications and cryptography because of its hyperchaos and feasibility. Further studies in this research direction will be presented in future works.

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